Genetic Algorithms for Distributed MIMO Radar Antenna Position Optimization

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Abstract—In multiple-input multiple-output (MIMO) radars, carrier frequency, signal bandwidth and antenna geometry have a deep impact on the ambiguity function (AF). In particular for systems employing widely separated antennas, sidelobes and azimuth ambiguities may appear in the MIMO-AF, potentially leading to a degradation of the system detection and localization capabilities.

The aim of this paper is to optimize antenna positions along the MIMO baseline using genetic algorithms (GAs). Key performance indicators (KPIs), such as the peak-to-maximum and peak-toaverage sidelobe ratios, as well as the range and cross-range resolutions are investigated as potential optimization criteria.

As a practical study case, a MIMO radar working the in Xband is simulated. The system employs two transmitters (TXs) and four receivers (RXs), with free-located TX-RX antenna pairs. The analysis is conducted for a point-like target at different positions. The optimization is carried out by means of the GAbased function library available in MATLAB[©], selecting both single and multiple KPIs as optimization criteria. In this latter case, the advantage is the optimization of more KPIs at the same time, however at the expense of a larger computation time.

Index Terms—MIMO Radar, Ambiguity Function, Antenna Position Optimization, Genetic Algorithm, Key Performance Indicators.

I. INTRODUCTION

The ever increasing demand of systems with superior resolution, stability and accuracy in high-precision civilian and industrial applications [1], [2] is pushing radars to become even more ubiquitous sensors [3].

In this context, the ceaseless progress in the miniaturization of electronic components, and the advancements in microwave

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photonics (MWP) techniques [4]–[6], are accelerating the realization of coherent multiple-input multiple-output (MIMO) radars, which were theorized almost twenty years ago [7] and that can be categorized in systems with co-located antennas [8] and systems with widely separated antennas [9].

Focusing on this second type of architecture, several parameters influence the overall peak-to-sidelobe ratio (PSR) of the MIMO ambiguity function (AF). As pointed out in [9], not only increasing the spatial information brings a reduction of the sidelobe level, but also the fractional bandwidth of the signal (i.e., the bandwidth used in transmission with respect to the carrier frequency) and the antenna geometry have a deep impact on the overall PSR [10], [11].

Thus, the concept of *information diversity* was introduced in [12], with the aim of understanding the system effectiveness in detecting and resolving closely spaced targets, as well as in suppressing sidelobes in the MIMO-AF at the varying of system geometry and frequency parameters.

For this reason, performance metrics were proposed and evaluated to characterize the effects that information diversity has on MIMO radars with widely separated antennas [13]. Among these, the peak-to-maximum and peak-to-average sidelobe ratios, as well as the range and cross-range resolutions were proposed as performance metrics.

On the other hand, these KPIs could reveal precious also for optimizing the TX and RX antenna positions along the MIMO baseline. Generalizing the study conducted in [14], this paper attempts at outlining the most suitable criteria for optimizing the antenna positions using genetic algorithms (GAs).

Previous works on using GA-based techniques for antenna position optimization in sparse arrays and MIMO radars were presented in [15] and [16], respectively. However, in [16], dealing with co-located MIMO radars, and in [14], dealing with distributed MIMO radars, the optimization criterion consisted in the minimization of the peak sidelobe level (PSL).

In this paper, the analysis is conducted for a point-like

static target at different positions. The optimization is carried out by means of the GA-based function library available in MATLAB©, selecting both single and multiple KPIs as potential optimization criteria.

II. MULTIPLE-INPUT MULTIPLE-OUTPUT RADARS

A coherent MIMO radar can employ M TX and N RX radar front-ends, not necessarily co-located. The front-ends are denoted with TX_m and RX_n , being $m = 1, \ldots, M$ and $n = 1, \ldots, N$, with $M \neq N$. For generality, TX_m can operate at L different radio frequencies (RFs).

A. MIMO Signal Model

Let $s_m^{(l)}(t)$ be the low-pass equivalent of the signal transmitted by TX_m at the l-th RF carrier, such that its waveform envelope has unit energy. The $M \times N$ antennas simultaneously illuminate K scatterers P_k having coordinates (x_k, y_k) , with $k = 1, \ldots, K$. These latter can belong to a single target or to multiple targets.

The time delay $\tau_{m,n}^{(k)}$ associated to the distance of P_k with respect to TX_m and RX_n is evaluated as follows:

$$
\tau_{m,n}^{(k)} = \frac{1}{c} \left[\mathbf{d} \left(T X_m, P_k \right) + \mathbf{d} \left(P_k, R X_n \right) \right],\tag{1}
$$

where c is the speed of light, and $d(A, B)$ represents the Euclidean distance between the generic points $A \equiv (x_a, y_a)$ and $B \equiv (x_b, y_b)$ in the 2D space:

$$
\mathbf{d}(A,B) = \sqrt{(x_a - x_b)^2 + (y_a - y_b)^2}.
$$
 (2)

In reception, the resulting $M \times N \times L$ individual *virtual channels* are separated for data processing. Thus, the low-pass equivalent of the signal received by RX_n can be written as in [9]:

$$
r_{m,n}^{(l)}(t) = \sum_{k=1}^{K} \zeta_{m,n}^{(k,l)} s_m^{(l)} \left[t - \tau_{m,n}^{(k)} \right] e^{j\varphi_{m,n}^{(l)}(t)} + w_{m,n}^{(l)}(t). \tag{3}
$$

Here, $\zeta_{m,n}^{(k,l)}$ denotes the complex amplitude of the received signal contribution due to the k-th scatterer, whereas $\varphi_{m,n}^{(l)}(t)$ accounts for the overall phase shift of the virtual channel.

It is worth noticing that the analysis of such shifts is out of the scope of this work. However, a model of phase noise induced by RF signal distribution through optical fiber links was presented and its impact evaluated in [17]. Additional considerations about the total angular jitter introduced by the system architecture can be found in [18]. Moreover, phase terms due to Doppler shifts are to be considered yet.

In eq. (3), the term $w_n^{(l)}(t)$ represents the overall clutterplus-noise contribution to the received signal, and, for simplicity, it is modelled as additive white Gaussian noise (AWGN) stochastic process. Finally, the terms $\zeta_{m,n}^{(k,l)}$ and $\tau_{m,n}^{(k)}$, with this latter described in eq. (1), depend on the bistatic geometry underlying the radar front-ends TX_m and RX_n , and the scatterer P_k over the *l*-th frequency channel $f_{RF}^{(l)}$:

$$
\zeta_{m,n}^{(k,l)} = \frac{1}{D_{m,k}^{TX} D_{n,k}^{RX}} \sqrt{\frac{P_m^{(l)} G_m^{(l)} A_n^{(l)} \sigma_{m,n}^{(k,l)}}{\left(4\pi\right)^3 k_B B_n^{(l)} T_n^{(l)}}},\tag{4}
$$

where $D_{m,k}^{TX} = \mathbf{d} (TX_m, P_k)$ and $D_{n,k}^{RX} = \mathbf{d} (P_k, RX_n)$.

In eq. (4), $P_m^{(l)}$ and $G_m^{(l)}$ are respectively the transmitted power and antenna gain at TX_m for the *l*-th waveform, $A_n^{(l)}$ is the effective area of the RX_n antenna for the l-th RF carrier $f_{RF}^{(l)}$, $\sigma_{m,n}^{(k,l)}$ is the bistatic radar cross section (RCS) of scatterer P_k observed by TX_m and RX_n , k_B is the Boltzmann's constant, $B_n^{(l)}$ is the noise bandwidth (BW), $T_n^{(l)}$ is the noise temperature at RX_n .

B. MIMO Radar Ambiguity Function

Let Θ_k be the k-th vector of parameters to be estimated. Under the assumption that scatterers do not interfere with each other, Θ_k , which for simplicity consists in the generic target position (x_k, y_k) in the 2D space, can be the determined from the maximum likelihood (ML) estimate evaluated from all the $M \times N \times L$ available virtual channels. As described in [9], the ML estimate can be obtained in two ways via the MIMO-AF.

The first way of calculating the MIMO-AF is based on a *non-coherent MIMO processing* approach, involving only the amplitude of the received signals. Instead, the second way is based on a *coherent MIMO processing* approach, taking into account also the phase of the received signals:

$$
\text{MIMO-AF}_c\left(\mathbf{\Theta}_k\right) \propto \left| \sum_{m=1}^M \sum_{n=1}^N \sum_{l=1}^L \varepsilon_{m,n}^{(k,l)} \cdot \Psi_{m,n}^{(k,l)} \left[t, \tau_{m,n}^{(k,l)}\right] \right|^2, \tag{5}
$$

where:

$$
\Psi_{m,n}^{(k,l)}\left[t,\tau_{m,n}^{(k,l)}\right] = \int_{-\infty}^{+\infty} r_{m,n}^{(k,l)}(t)s_m^{(l)*}\left[t-\tau_{m,n}^{(k,l)}\right]dt\tag{6}
$$

represents the cross-correlation between $r_{m,n}^{(k,l)}(t)$ and $s_m^{(l)}(t)$. To obtain an overall picture of the monitored area, the MIMO-AF in eq. (5) is evaluated for each point (x, y) in the observation space. Finally, the exponential term:

$$
\varepsilon_{m,n}^{(k,l)} = e^{-j2\pi f_{RF}^{(l)} \tau_{m,n}^{(k,l)}} \tag{7}
$$

depends on the l-th RF carrier and on the underlying bistatic geometry among TX_m , RX_n and the scattering element P_k . After this phase compensation, the complex correlation contributions in eq. (6) can be summed together coherently, as described by eq. (5).

C. Key Performance Indicators

In technical terms, KPIs are parameters that quantify the performance of a system. In this paper, the relevant KPIs proposed in [12], identified for evaluating the performance of a MIMO radar with widely separated antennas, act as the criteria used by the genetic algorithm for the optimization of the antenna positions.

The GA, that will be presented in Section III, exploits the following KPIs both in an individual or joint manner (i.e., trying maximize two KPIs at the same time):

- Peak-to-Maximum Sidelobe Ratio (PMSR),
- Peak-to-Average Sidelobe Ratio (PASR),
- Range resolution (ΔR) of the mainlobe,
- Cross-range resolution $(\Delta X R)$ of the mainlobe.

III. GENETIC ALGORITHMS FOR ANTENNA POSITION **OPTIMIZATION**

In MIMO radars, the TX and RX antennas create a $M \times N$ equivalent virtual array given by the Kronecker product of the M -ary TX array with the N -ary RX array [19]. This way, it is like if the RX array were repeated for every transmitting antenna. However, this can lead to redundancy in the virtual array elements placement [20]. Therefore, it is necessary a robust procedure for optimizing the distribution of the MIMO radar antennas along the baseline.

Genetic algorithms represent a class of algorithms for solving both constrained and unconstrained optimization problems [21]–[23]. They follow an approach similar to the one driving biological evolution and natural selection. They can be applied to solve optimization problems in which the objective function is discontinuous, non-differentiable, stochastic, or highly nonlinear. Finally, they help optimizing the attributes of a given set of objects called *population*. The high-level flow diagram of the system optimization procedure based on genetic algorithm is sketched in Fig. 1, whereas the main differences with the statistical-based approach are summarized in Table I.

Fig. 1. High-level flow diagram of optimization based on genetic algorithm.

The core of the algorithm is the *fitness function*. In principle, the fitness function can be any of the already defined KPIs. Population reproduction can be obtained by mutations, i.e., statistical perturbations of the peripherals number and position. From a detection and estimation perspective, the main interest is in minimizing the sidelobes, understanding how they behave changing antennas disposition and number. Instead, from a target localization perspective, the aim is to minimize the system resolution, both in range and cross-range domains.

Since Matlab already has a set of functions related to genetic algorithms [24], this operation is performed using the ga and

Statistical Approach	Genetic Approach			
The algorithm generates a single	Generates a population of points			
point at each iteration. The	at each iteration. The best point in			
sequence of points approaches an	in the population approaches an			
optimal solution.	optimal solution.			
Selects the next point in the	Selects the next population by			
sequence by a deterministic	computation which uses random			
computation.	number generators.			
Typically converges quickly to a	Typically takes many function			
local solution.	evaluations to converge. May or			
	may not converge to a local or			
	global minimum.			
TABLE I				

STATISTICAL VS GENETIC APPROACH

gamultiobj built-in functions. The first is used when the fitness function is represented by only one optimization criterion, whereas the second is used when multiple optimization criteria are used.

IV. SIMULATION RESULTS

In the simulations, which replicate the in-door experimental scenario described in [25], the following setup parameters are considered for the coherent MIMO radar system:

- $M = 2$ TXs and $N = 4$ RXs over a 3 m baseline;
- Target distance from the baseline center equal to 3 and 30 m, for sudy case A and B, respectively;
- Frequency $f_{RF} = 10$ GHz, with signal bandwidth $B = 1$ GHz (i.e., 1/10 fractional bandwidth).

The optimization procedure is carried out fixing one TX and one RX at the two extremes of the baseline. This allows to exploit the whole baseline extent and, thus, to achieve the maximum nominal azimuth resolution imposed by the baseline length given the RF frequency.

Seven criteria are considered as fitness functions for the GA. Four of them consist in optimizing separately PMSR, PASR, ΔR and $\Delta X R$. They are indicated with criteria no. 1, 2, 4 and 5, respectively. Instead, the last three criteria consist in the joint optimization of two KPIs simultaneously: PMSR and PASR, ΔR and ΔXR , PMSR and ΔXR are indicated with criteria no. 3, 6 and 7, respectively.

A. Analysis of Study Case A

The MIMO-AFs corresponding to the seven optimized antenna configurations are depicted in Fig. 2, 3 and 4. In particular, in Fig. 2 results of the optimization of PMSR, PASR and joint PMSR and PASR are shown. In Fig. 3, results of the optimization of ΔR , ΔXR and joint ΔR and ΔXR are shown. Finally, in Fig. 4, results of the joint optimization of PMSR and ΔXR are shown. For completeness, the KPIs measured on the seven optimized MIMO configurations are summarized in Table II.

In general, regardless of the optimization criteria, the second TX is placed at the opposite side of the baseline with respect to the first one. Conversely, the RXs are more distributed along the baseline, especially when the optimization is done to reduce the MIMO-AF sidelobes (i.e., PMSR, PASR). In this

Fig. 2. MIMO baseline optimization based on the maximization of the peak-to-sidelobe ratios. No. 1) maximization of PMSR (left column): a) resulting coherent MIMO output, b) optimized MIMO baseline; No. 2) maximization of PASR (center column): c) resulting coherent MIMO output, d) optimized MIMO baseline; No. 3) joint maximization of PMSR and PASR (right column): e) resulting coherent MIMO output, f) optimized MIMO baseline.

Fig. 3. MIMO baseline optimization based on the minimization of system resolution parameters. No. 4) minimization of ∆R (left column): a) resulting coherent MIMO output, b) optimized MIMO baseline; No. 5) minimization of $\Delta X R$ (center column): c) resulting coherent MIMO output, d) optimized MIMO baseline; No. 6) joint minimization of ΔR and $\Delta X R$ (right column): e) resulting coherent MIMO output, f) optimized MIMO baseline.

case, the maximization of PMSR is more effective than the maximization of PASR.

Instead, minimization of ΔR and/or ΔXR could result dangerous if they are the only optimization criteria, because the resulting sidelobe level may become too large. Thus, if minimization of resolution is sought, it should be always accompanied by minimization of sidelobes, too (see optimization criterion no. 7). Moreover, joint criteria have the advantage of optimizing more KPIs at the same time, even if at the expense of a larger computation time.

B. Analysis of Study Case B

For conciseness, the resulting MIMO-AFs and corresponding array configurations are not shown. However, the analysis of KPIs after antenna position optimization when the target is

No.	Opt. Criterion	PMSR [dB]	$PASR$ [dB]	ΔR [m]	$\Delta X R$ [m]
1°	PMSR	5.3080	15.2199	0.1140	0.0150
$\mathbf{2}$	PASR	1.3556	16.2447	0.1320	0.0135
3)	PMSR & PASR	5.3042	15.4150	0.1260	0.0150
4	ΔR	2.9923	14.8281	0.1020	0.0150
5)	\triangle <i>XR</i>	0.2862	13.7568	0.1500	0.0135
6)	$\Delta R \& \Delta X R$	1.3683	14.9017	0.01020	0.0135
	PMSR & ΔR	5.3144	15.0889	0.01260	0.0135

TABLE II PERFORMANCE RESULTS OF THE OPTIMIZATION CRITERIA FOR A TARGET AT 3 M (STUDY CASE A)

at 30 m from the centre of the MIMO baseline are summarized in Table III. When the target position changes in the space, results demonstrate that similar conclusions can be drawn

Fig. 4. MIMO baseline optimization based on criterion No. 7) joint maximization of PMSR and $\Delta X\overline{R}$: a) resulting coherent MIMO output, b) optimized MIMO baseline.

No.	Opt. Criterion	PMSR [dB]	PASR [dB]	ΔR [m]	ΔXR [m]
	PMSR	13.1291	21.4410	0.1380	0.1320
$^{2)}$	PASR	13.1029	23.0628	0.1380	0.1500
3)	PMSR & PASR	13.1293	22.4838	0.1380	0.1410
4	ΔR	11.8667	20.1933	0.1380	0.1500
5)	\triangle <i>XR</i>	0.3600	10.1010	0.1380	0.1155
6)	$\Delta R \& \Delta X R$	1.0931	12.7024	0.1380	0.1150
7	PMSR & ΔXR	13.0308	21.2694	0.1380	0.1305

TABLE III PERFORMANCE RESULTS OF THE OPTIMIZATION CRITERIA FOR A TARGET AT 30 M (STUDY CASE B)

on the effectiveness of KPIs to be chosen. Even if only two points in the monitored space have been considered, it is possible to conclude that every target position leads to a different optimized MIMO radar configuration. Thus, the optimum MIMO radar configuration should be retrieved by *averaging* the results over the whole area.

An additional prosecution of this work will concern the analysis of non-static targets, which entail phase compensation due to target-sensor relative velocity, as well as the optimization of MIMO radar baselines in close-to-reality scenarios (e.g., coastal-based systems, swarms of drones). Finally, due to the randomness in the optimization process, Monte Carlo simulations will be considered for obtaining a more accurate performance analysis of the algorithm. The aforementioned issues are currently object of ongoing research.

V. CONCLUSION

In this paper, the optimization of antenna positions in a multiple-input multiple-output (MIMO) radar using genetic algorithms (GAs) has been presented. Key performance indicators (KPIs) measured on the MIMO ambiguity function, such as the peak-to-maximum and peak-to-average sidelobe ratios, respectively PMSR and PASR, as well as the range and cross-range resolutions have been investigated as potential optimization criteria. The optimization has been carried out by means of the GA-based function library available in MATLAB©, selecting both single and multiple KPIs as optimization criteria.

The maximization of PMSR is more effective than the maximization of PASR. Minimization of range and/or crossrange resolutions could result dangerous if they are the only optimization criteria. Thus, if minimization of resolution is sought, it should be always accompanied by minimization of sidelobes. Even if only two points in the monitored space have been considered, it is possible to conclude that every target position leads to a different optimized MIMO radar configuration. Thus, the optimum MIMO radar configuration could be retrieved by averaging the results over the whole area.

Additional prosecution of this work will concern the analysis of non-static targets, as well as the optimization of MIMO radar baselines in close-to-reality scenarios (e.g., coastal-based systems, swarms of drones). Finally, Monte Carlo simulations will be considered for obtaining a more accurate performance analysis of the algorithm, thus overcoming the solution randomness issue in the GA-based optimization process.

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